

Radio Design 201 Microwave Circuits and Antennas

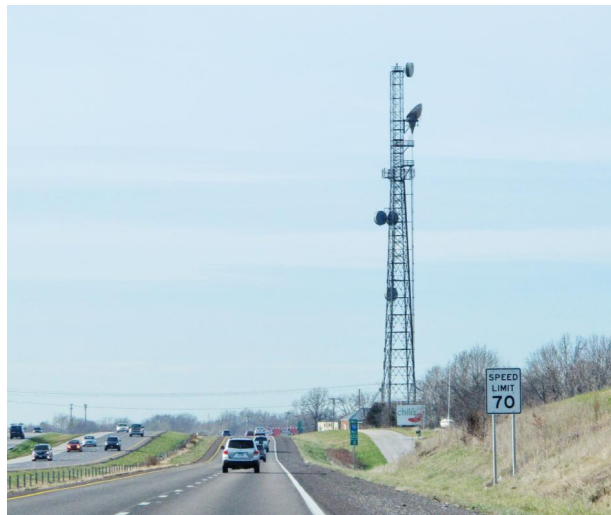
Episode 5 - Gain, Path Loss, Spectrums, and Noise

Slides downloaded from: <https://ecefiles.org/>

Companion videos at: <https://www.youtube.com/playlist?list=PL9Ox3wpmB0kpOb7NdpHAK-jDpsZm3Xedx>

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This episode of Radio Design 201 (Microwave Circuits and Antennas) focuses on fundamentals of communication links, including how to calculate received signal level, and the role of noise in a system's ability to detect and demodulate the signal. At microwave frequencies, the antennas used often have high directivity and gain, allowing for very long-range communication. Signal path-loss is considered and the use and effects of path-loss exponents from 2 to 5 is explained. The video concludes with demonstrations of spectrum analyzers, including a TinySA Ultra and a comparison of the TinySA to an older high-performance Hewlett Packard unit.

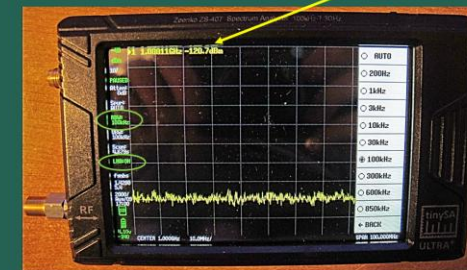


Spectrum Analyzer Noise Floor

$$P_n = -174 \text{ dBm} + 10 \log(B) + NF_{rcvr}$$

TinySA Ultra spectrum analyzer with RBW = 100 kHz, Preamp ON, (NF = 3 dB)

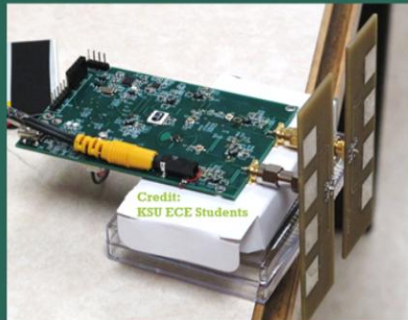
$$P_n = -174 \text{ dBm} + 10 \log(1E5) \text{ dB} + 3 \text{ dB} = -121 \text{ dBm}$$



Radio Design 201 #5
Microwave Circuits and Antennas
Gain, Path Loss, Spectrums, and Noise



Radio Design 201 #1 Microwave Circuits and Antennas



MegawattKS - YouTube



<https://ecefiles.org>

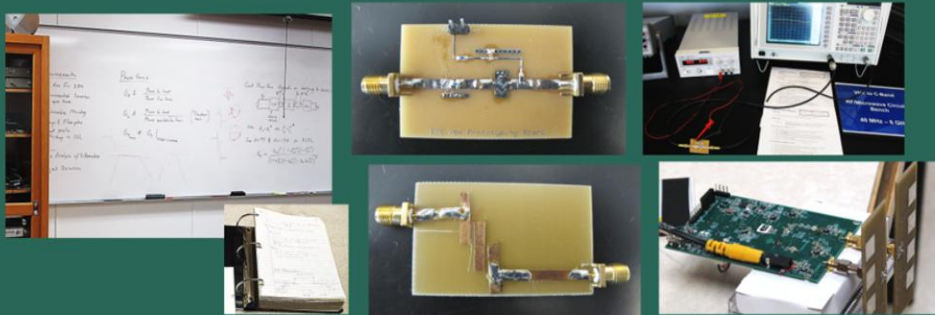
Radio Design 201 #2 Microwave Circuits and Antennas Project & Product Examples



MegawattKS - YouTube

<https://ecefiles.org>

Radio Design 201 #3 Microwave Circuits and Antennas Project-oriented Learning



MegawattKS - YouTube

<https://ecefiles.org>

Radio Design 201 #4 Outfitting Microwave Labs On a Budget

Outfitting an RF/Microwave Lab on a Budget 2001 - 2020

Modern RF and microwave test equipment is awesome. It can provide accurate measurements to 20 GHz and above in relatively small, lightweight packages with bright color screens and many time-saving data-reduction options. But it is also extremely expensive. The photo at left is an HP8753E we bought from the refurbished market. Newer models sell for \$50K to \$100K depending on frequency range!

For those of us who must live within reasonable means, such instruments are usually outside our reach. However, there are ways to continue working in this business - if we are willing to give up on some of the fancy looks and features. Fortunately, this does NOT include giving up on basic performance (much). The picture at right is a 20 GHz network analyzer assembled from pieces bought for less than \$2K total, after scouting Ebay and used-line.com.

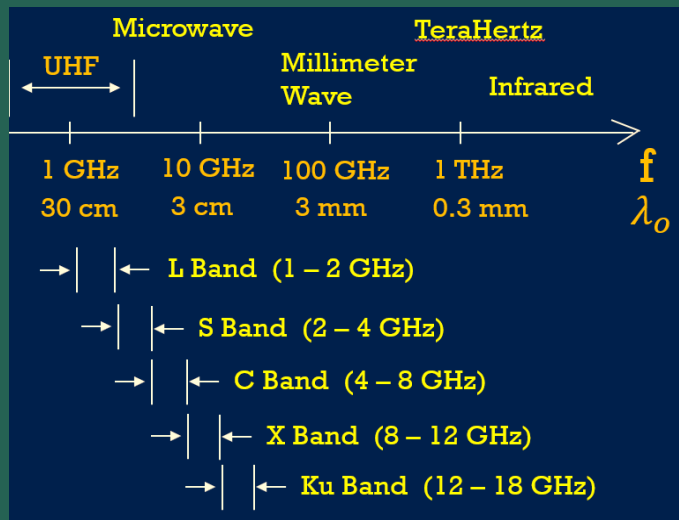


MegawattKS - YouTube

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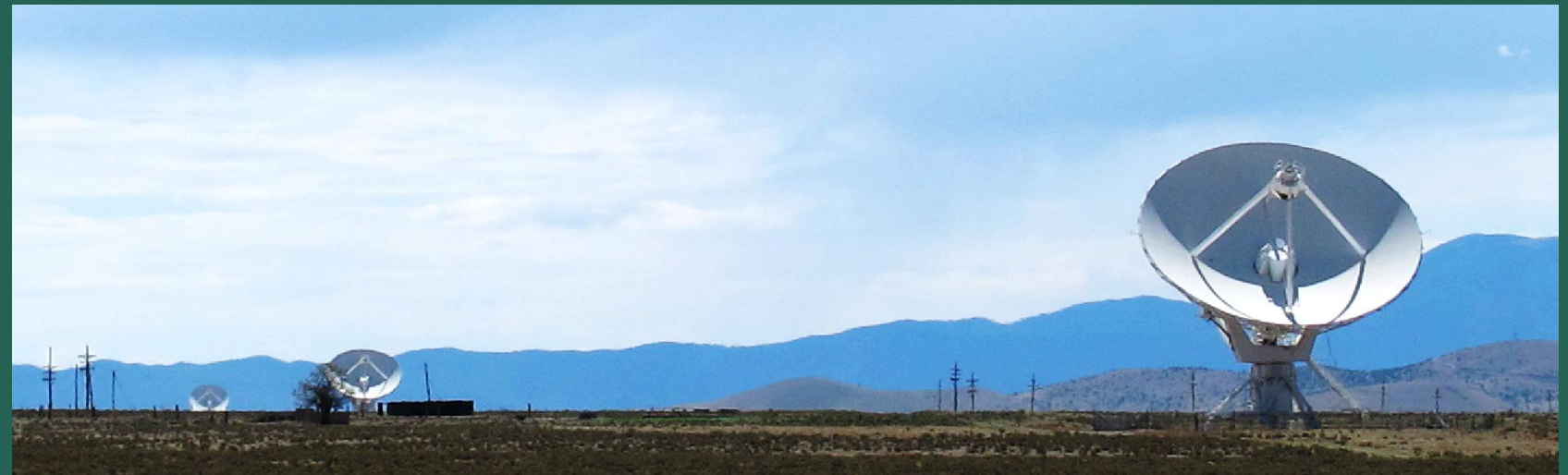


<https://ecefiles.org>



Why Use “Microwaves” ?

- Massive Amounts of Spectrum Available
- Highly Directional Antennas Possible
- Very Long Range (for line-of-sight paths)



Today's "Lecture" Topics


Announcements

Welcome to Episode 5

See Episodes 1 through 4 for

- *Definitions of "microwave",*
- *Example semester projects, and*
- *Day 1 lecture material*
- *Free and low-cost software and test equipment*

Today's Topics

 *Finding received power levels in free-space*

Noise and Minimum Discernable Signal (MDS) levels

Spectrum Analyzer Usage and Demos

Today's Episode in Context

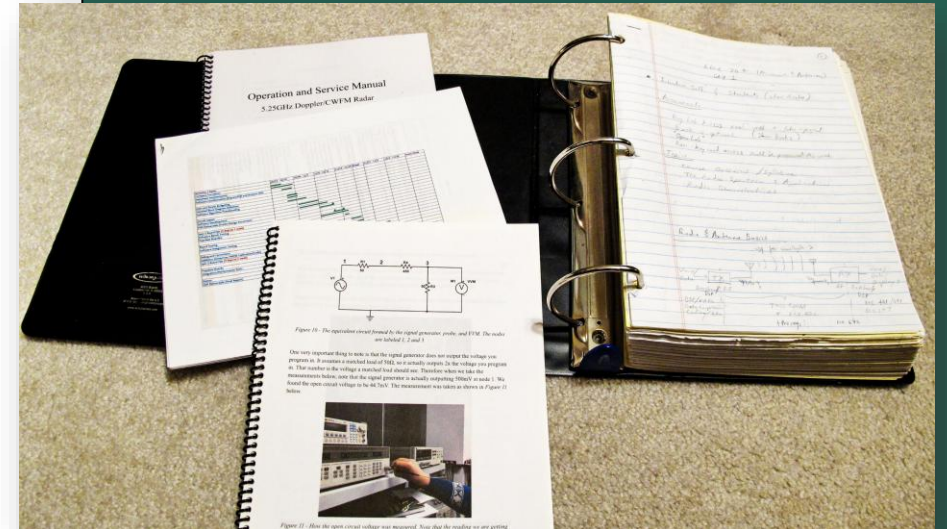
In Episode 3
(p/o 4-part introduction)

This Episode

ECE 764 Spring 2017 Schedule

(Tentative)

| Week | Material | Assignments |
|------|---|--|
| 1 | Introduction to comm systems, frequencies, and wavelengths <u>Calculating and measuring received signal strengths</u> | Verify you have lab access Lab 1 assigned |
| 2 | EM theory review and demos Basic antenna theory | Purchase lab kit |
| 3 | Basic antenna types: dipoles, monopoles, loops Antenna characteristics: polarization, patterns, and impedance | Lab 1 due |
| 4 | Measurement of antenna patterns and signal reflection/transmission Transmission lines and reflection coefficient | Lab 2 assigned |
| 5 | Introduction to Smith Charts, plotting of impedance/admittance Simulation with Agilent ADS | |
| 6 | Lambda/4 lines and impedance transformation with microstrip Microstrip components and coupled-line microwave filters | Lab 2 due |
| 7 | PC board design, layout, and parasitics Two-port circuits and S-parameters | Lab 3 assigned |
| 8 | Circuit analysis and simulation with S-parameters Amplifier insertion gain, matching, stability, noise figure, and compression specs | Lab 3 due |
| 9 | Project discussion (including needed baseband, ADC and DSP functions) Basic mixer concepts and hybrid-ring microstrip implementations | Project assigned |
| 10 | Voltage controlled oscillators Frequency multipliers | Prototype design/layouts due |
| 11 | Project discussions (pads, RF connector test points, test planning) | Probotype boards back |
| 12 | Frequency synthsizers and phase-locked loops | |
| 13 | Project testing/debugging discussions | Test, debug, respin boards Submit final board layouts |
| 14 | Analysis and design of directional antennas: Array theory Dish antennas and antenna feeds: e.g. helix, horn, and patch antennas Antenna simulations with ADS, EZNEC and other tools | |
| 15 | Additional radar and wireless performance issues | |

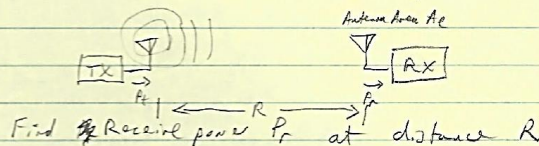


Today's Lecture Notes

Transmit & Receive Power Levels

Problem

Consider a transmitter TX with output power P_t
 And a receiver RX with antenna of size A_e (effective aperture)



Find the Receive power P_r at distance R

Solution:

Assume TX antenna is isotropic

Propagation medium is free space (no obstructions/etc)

Then, from power conservation:

Power density at distance R is

$$P = \frac{P_t}{4\pi R^2} \left(\frac{W}{m^2} \right)$$

Area of sphere around TX antenna

And

$$P_r = P A_e$$

$$= \frac{P_t}{4\pi R^2} A_e$$

Note: If R varies from 1m to 10km (10^4)

Then P_r varies by 10^8 Eight orders of magnitude

⇒ Watts or mW → μW to pW

P_r therefore commonly measured in log terms w/ dB units

expressed by

$$P_r |_{dBm} = 10 \log \frac{P_r}{1mW}$$

↑
dB relative to 1mW

| P | dBm |
|-----|------|
| 1W | +30 |
| 1mW | 0 |
| 1μW | -30 |
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| 1pW | -90 |
| 1fW | -120 |

Minimum Discernable Signal

Weakest power receivable is called MDS sensitivity

Limited by "noise floor"

$$P_n = kTB$$

← Temp in Kelvin
 ← BW of receiver's "detector" in Hz
 Boltzmann's Constant = $1.38 \times 10^{-23} \frac{W}{K \cdot Hz}$

Example $T = 290K$ (room temperature)

$B = 10KHz$ (modulation & receiver dependent)

$$\Rightarrow P_n = MDS = 4 \times 10^{-17} \text{ watts}$$

$$= 4 \times 10^{-14} \text{ mW}$$

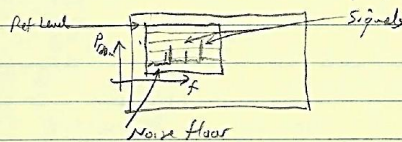
OR

$$P_n |_{dBm} = -134 \text{ dBm}$$

NOTE: Effective P_n & MDS of RX is higher by X dB
 where $X = \text{"Noise Figure"}$

Spectrum Analyzers

Radio receiver with power vs frequency readout display



Typical sensitivity may be 10 to 30 dB worse than theoretical value above, depending on settings

See handout / KSOC for more info

Spectrum Analyzer Demo (Pass out cheat sheets!)

Freq Sweep Settings: Start/stop for wide range
 Center/Spn for closer inspection
 Freq accuracy limited by RBW, Span/# points taken

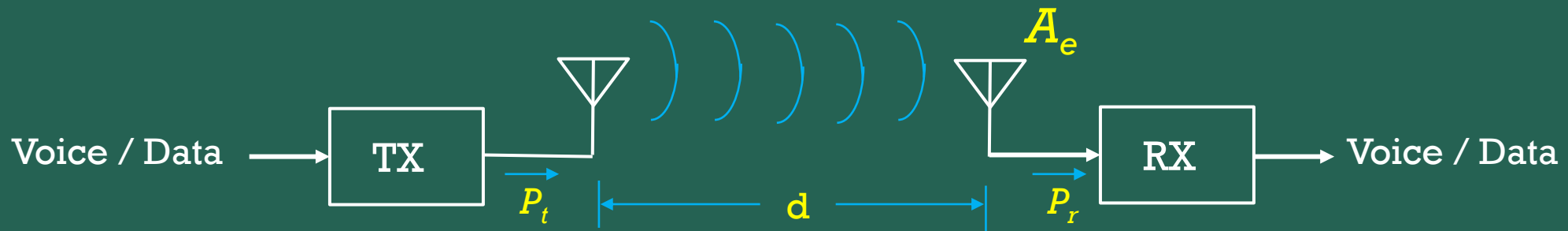
Amplitude Settings: Scale typically 10 dB/div vertical
 Ref Level may default to 0dB
 Need higher for large power
 Damage Level ≈ +20dBm!
 Need lower for weak signal observation

RBW & VBW: RBW is detector BW
 Decrease to decrease Noise floor
 (also lower Ref Level enable preamp decrease atten (but NOT below 10dB!))

Additional Functions: Marker, Peak select, Mkr → CF
 F count
Preset

Transmit & Receive Power Levels

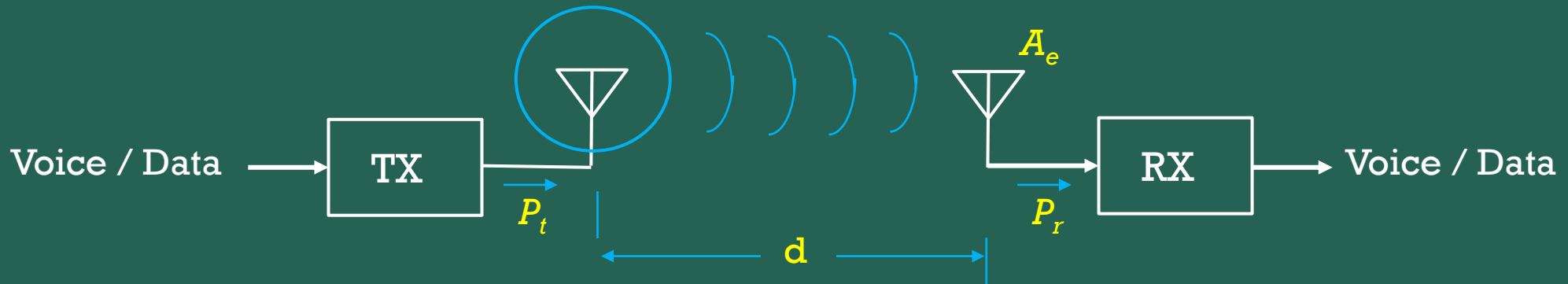
Consider a transmitter **TX** with output power P_t Watts
And a **RX** with antenna area (“effective aperture”) A_e m^2



Problem:

Find the received power P_r at distance d

Solution (basic / simplified)



Assume TX antenna is “isotropic” (non-directional)
RX antenna and feedline is well matched to RX input impedance
Propagation medium is free-space (no obstructions/reflections)

Then Power density $P_{density}$ at distance d is:

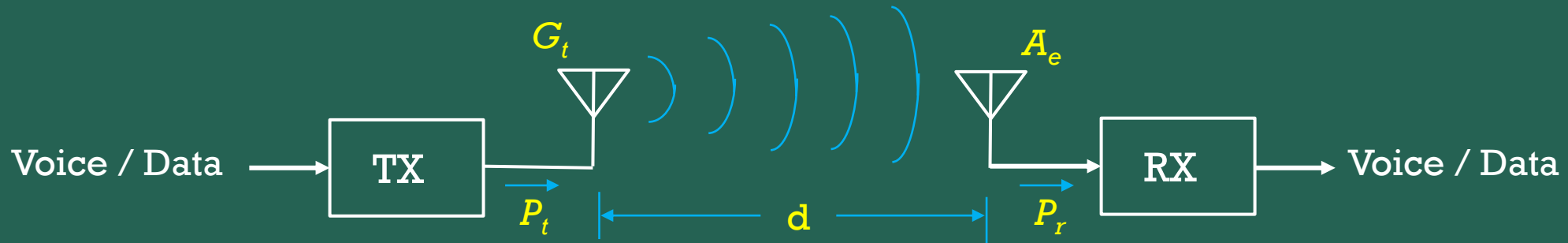
$$P_{density} = \frac{P_t}{4\pi d^2} \text{ Watts/m}^2$$

And received power P_r is:

$$P_r = P_{density} A_{eff}$$

Surface area of a sphere of radius d surrounding “isotropic” TX antenna

Antenna Gains and Path Loss Exponent



Real TX antennas have directivity and associated “gain” (G_t)

(and so do RX antennas ...)

Terrestrial propagation paths may have obstructions / reflections, so “path loss exponent” in denominator is often approximated as $n = 2$ to 5

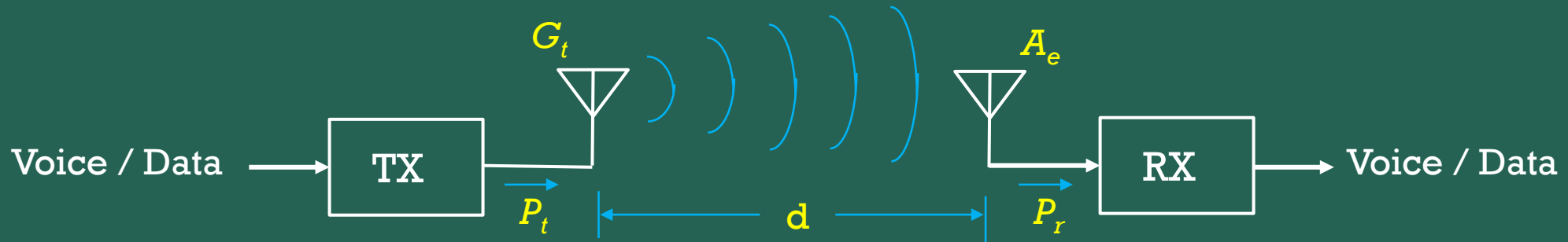
And P_r formula becomes:

$$P_r = \frac{P_t G_t}{4\pi d^n} A_e$$



See: [Antenna Briefs, Episode 7](#)

Important Notes:

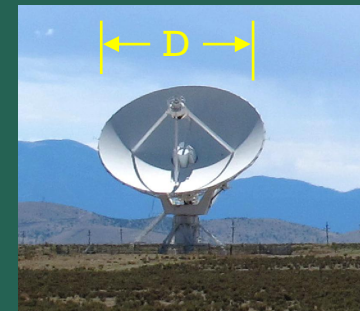


Basic (half-wave) dipole antennas have gain of

$$G \approx 1.6 \text{ (2 dBi)}, \text{ and } A_e \approx \frac{1}{\pi} \left(\frac{\lambda}{2}\right)^2$$

Dish antennas can have large effective apertures and very large gain at microwave frequencies

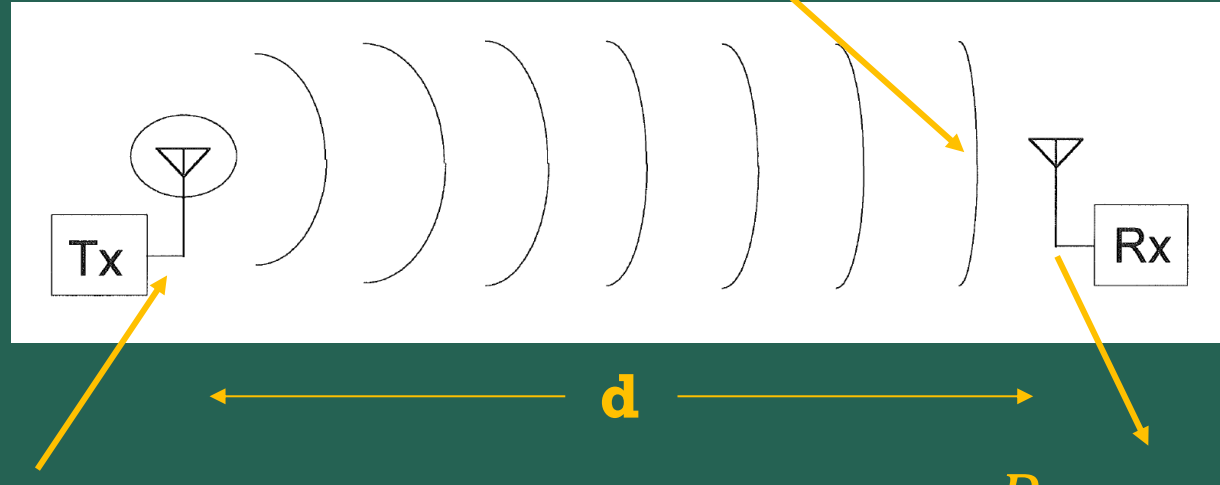
$$A_{eff} \approx \pi \left(\frac{D}{2}\right)^2 \text{ and } G = \frac{A_{eff}}{A_{ei}} = \frac{4\pi A_{eff}}{\lambda^2}$$



Example / Demo



$$P_{density} = \frac{P_t G_t}{4\pi d^n} \text{ Watts/m}^2$$



$$P_t = \frac{V_t^2}{R_{ant}} \text{ Watts}$$

Example: 1 V_{rms}, 50 Ohms
 => **Pt = 20 mW**

NOTE: NanoVNA actually puts out less ...

$n \approx 2$ here?
 Why?

$$P_r = P_{density} A_{eff} \text{ Watts}$$

Example: $G_t = 1.6$, $d = 0.5\text{m}$
 => $P_{density} = 10 \text{ mW/m}^2$
 Then with $A_{eff} \approx 1/3.14 (0.34\text{m})^2$
 => **Pr ≈ 0.4 mW**

(About $10 \log(0.4/20) = 17 \text{ dB}$ less than Pt)

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Spectrum Analyzer Usage and Demos

Power in dBm units

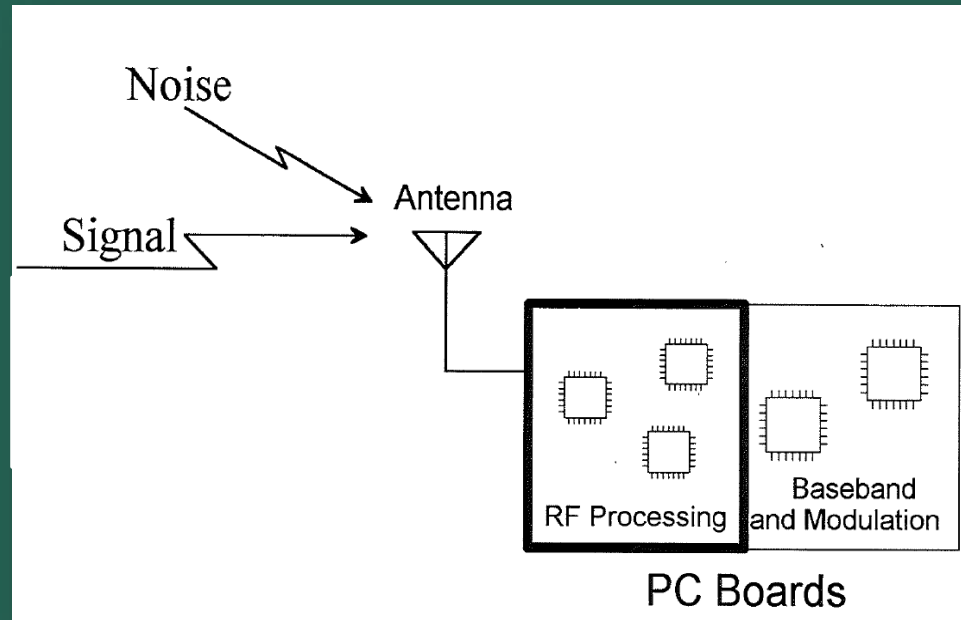
- Received power P_r is usually orders of magnitude lower than P_t
- If d varies from 1m to 10 km (about 6 miles), P_r decreases by a factor of 10^8 ! (e.g. from 0.1mW to 1 pW, for $n=2$. Much worse for $n=4$)
- So P_r often expressed in dB relative to 1mW, or “dBm”

$$P_r |_{\text{dBm}} \triangleq 10 \log \frac{P_r}{1\text{mW}}$$

| Power | dBm |
|--------|---------|
| 1 Watt | +30 |
| 1 mW | 0 |
| 1 uW | -30 |
| 1 nW | -60 |
| 1 pW | -90 😊 |
| 1 fW | -120 😐 |
| 1 aW | -150 ☹️ |

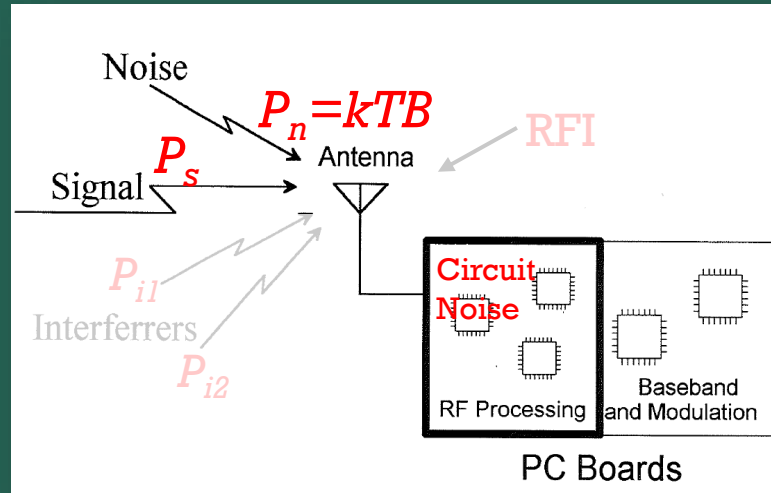
Minimum Discernable Signals

- Weakest power receivable often taken as the level of incoming noise
- Called the Minimum Discernable Signal level, or MDS
- Actual minimum depends on receiver design, modulation, coding, ...



Overall Receiver Sensitivity

See Radio Design 401, Episode 2 for details



$$\begin{aligned} P_{n \text{ dBm}} &= 10 \log \left(\frac{kTB}{0.001} \right) \\ &= 10 \log \left(\frac{kT}{0.001} \right) + 10 \log(B) \\ &\quad \underbrace{\hspace{10em}} \\ &= -174 \text{ dBm @ } T=290\text{K} \end{aligned}$$

$$P_n = -174 \text{ dBm} + 10 \log(B)$$

$$P_{s_min} = P_n + NF_{rcvr} + S/N_{min}$$

Example: TinySA Ultra spectrum analyzer with $B=100 \text{ kHz}$, $NF = 3 \text{ dB}$,

$$\Rightarrow P_n = -174 \text{ dBm} + 10 \log(1E5) \text{ dB} + 3 \text{ dB} = \underline{-121 \text{ dBm}}$$

Spectrum Analyzer Noise Floor

$$P_n = -174 \text{ dBm} + 10 \log(B) + NF_{rcvr}$$

TinySA Ultra spectrum analyzer with RBW = 100 kHz, Preamp ON, (NF = 3 dB)

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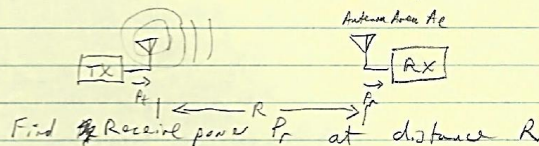
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Area of sphere around TX antenna

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 Limited by "noise floor"

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 BW of receiver's "detector" in Hz

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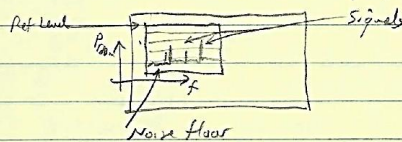
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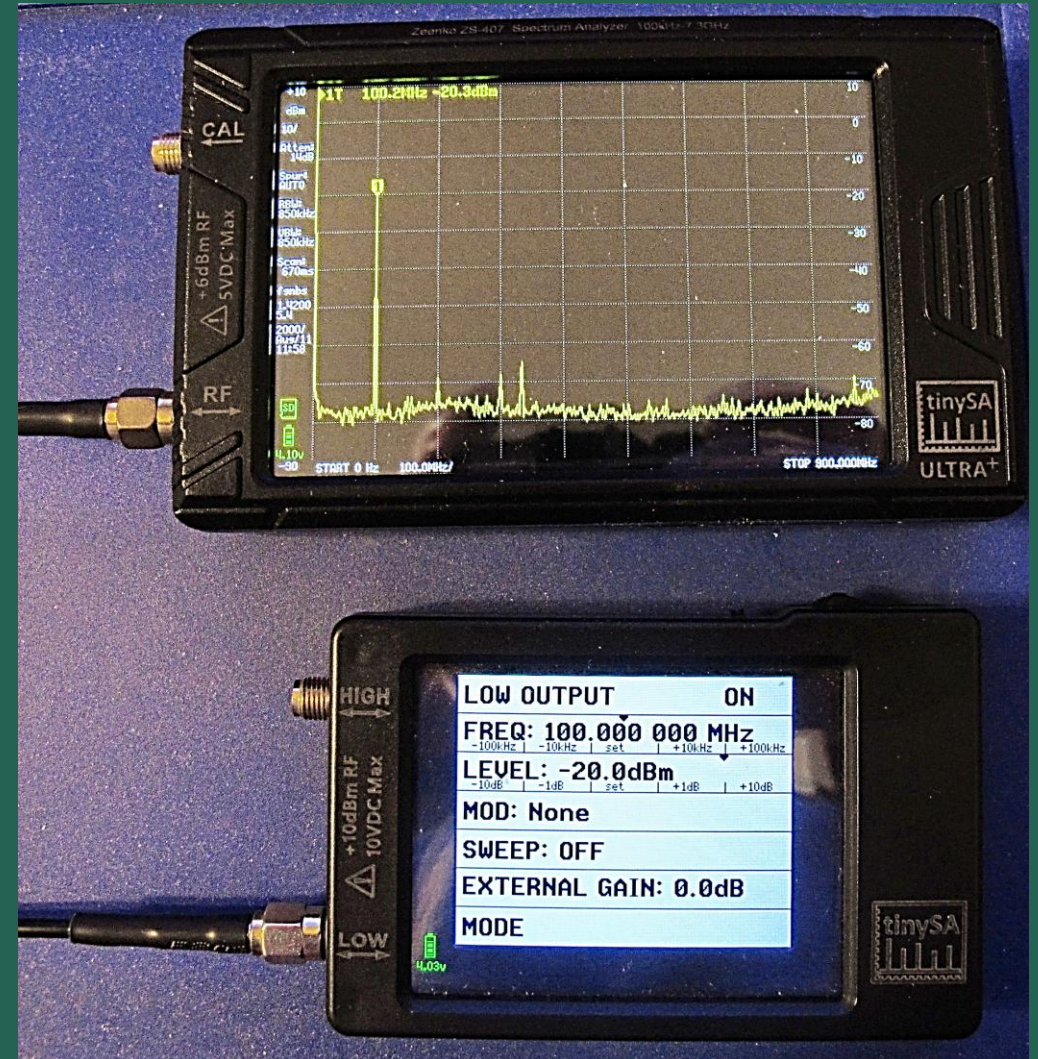
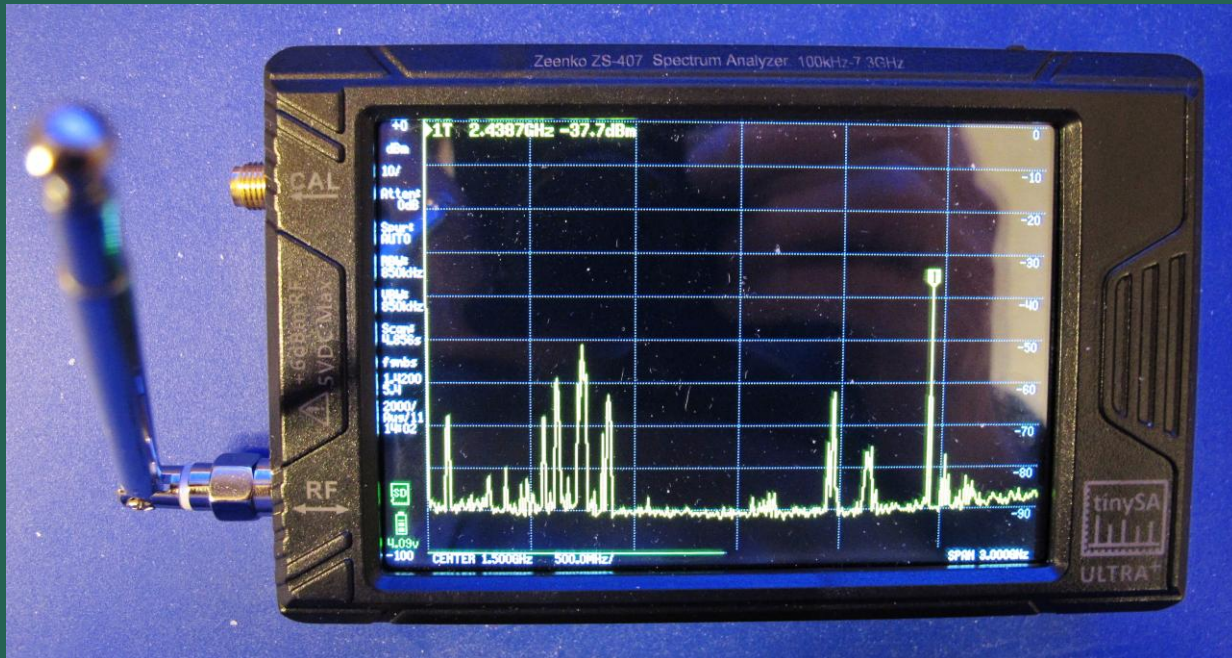
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Spectrum Analyzer Demos



*Thanks For
Watching !*

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