

Radio Design 101

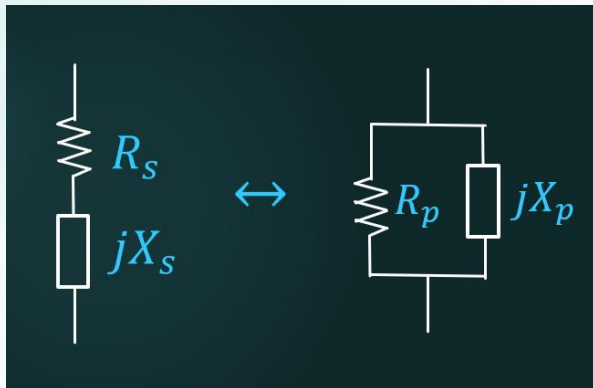
Appendix B - RF Impedance Conversions

Slides downloaded from: <https://ecefiles.org/rf-design/>

Companion video at: <https://www.youtube.com/watch?v=vO-AJjIX7a4>

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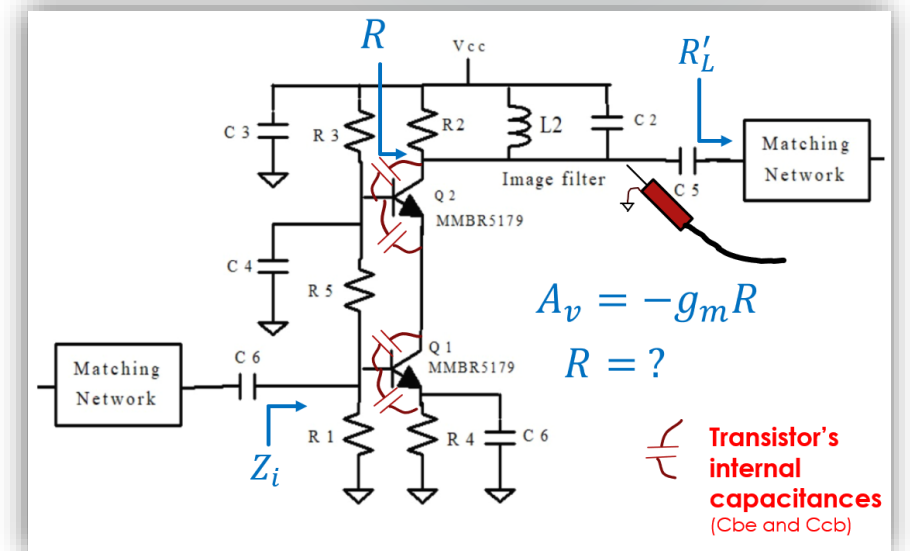
This video covers series to parallel impedance conversion, its use in matching networks, and in designing and measuring practical RF circuits. Example applications include understanding impedance values shown by RF test equipment and circuit simulations. In addition, an RF cascode amplifier is used to illustrate issues with real-world inductors, oscilloscope probes, and matching to complex impedance values.



$$R_p = (1 + q^2)R_s$$

$$X_p = \left(1 + \frac{1}{q^2}\right)X_s$$

where $q = \frac{X_s}{R_s} = \frac{R_p}{X_p}$

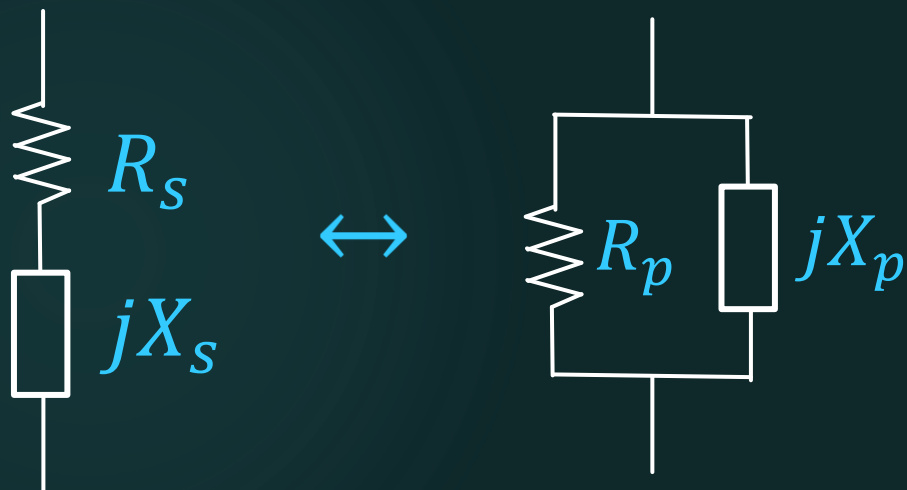


Radio Design 101

Appendix B

RF Impedance Conversions

for Matching, Amplifiers, Simulation, and Measurements

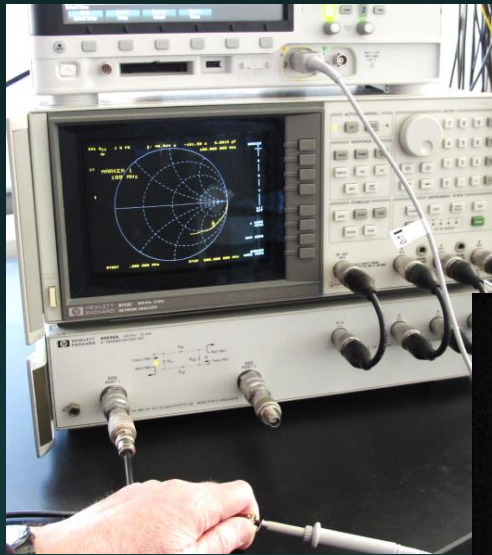


$$R_p = (1 + q^2)R_s$$

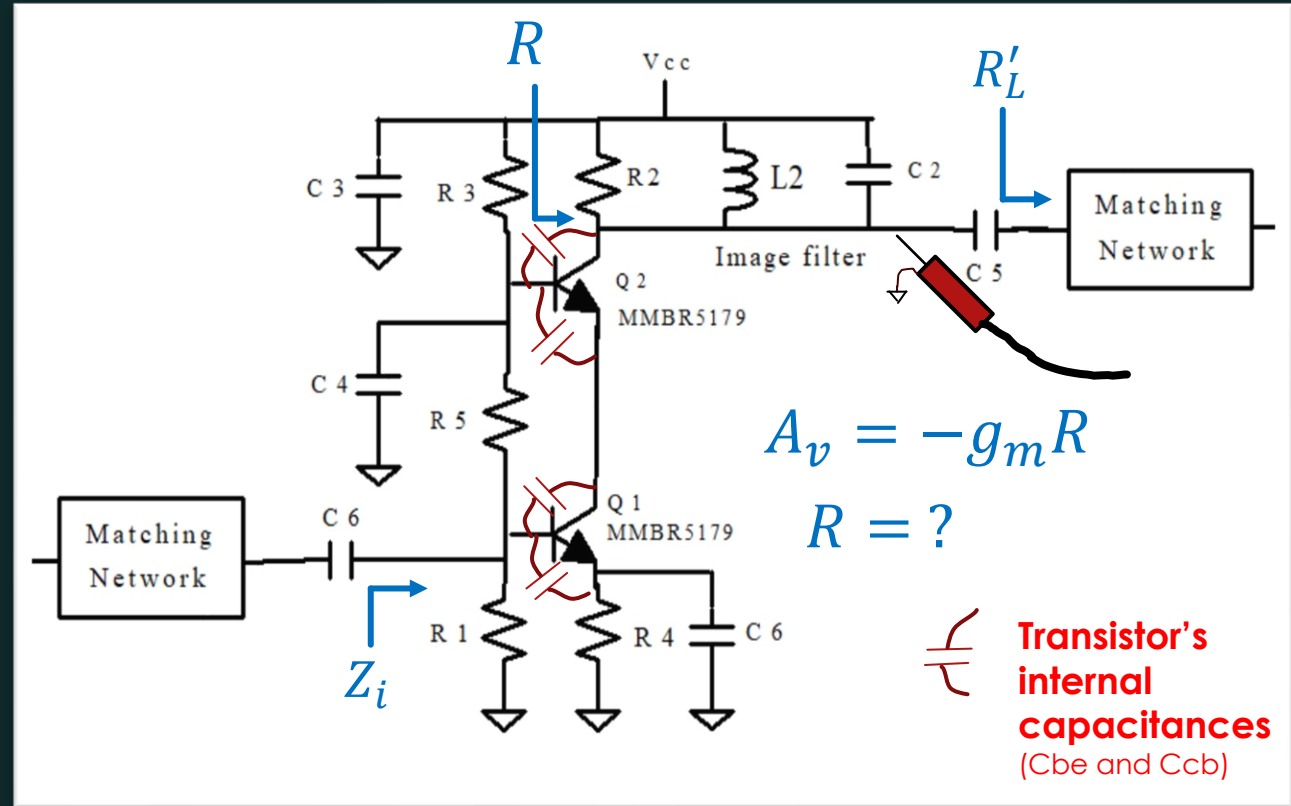
$$X_p = \left(1 + \frac{1}{q^2}\right)X_s$$

where $q = \frac{X_s}{R_s} = \frac{R_p}{X_p}$

Applications



Series to Parallel Impedance Conversion



Matching Networks & Amplifier Design

Lots more !

Background

Radio Design 101
MegawattKS
13 videos • 46,321 views • Last updated on Nov 22, 2023

New Radio Design 101 series
Abstracted from a senior-design University class

- 1. Radio Design 101 - Episode 1 - Transceivers and Filters - Part 1 (26:14)
- 2. Radio Design 101 - Episode 1 - Transceivers and Filters - Part 2 (30:43)
- 3. Radio Design 101 - Episode 2 - Impedance Matching - Part 1 (14:27)
- 4. Radio Design 101 - Episode 2 - Impedance Matching - Part 2 (27:43)
- 5. Radio Design 101 - Episode 3 - RF Amplifiers (50:39)
- 6. Radio Design 101 - RF Oscillators (Episode 4) (38:23)
- 7. Radio Design 101 - RF Mixers and Frequency Conversions - Episode 5, Part 1 (32:46)
- 8. Radio Design 101 - RF Mixers, Part 2 of Episode 5 (36:38)
- 9. Radio Design 101 - Finishing the Receiver (Episode 6) (35:05)

NanoVNA and Radio Frequency / Microwave Tech
MegawattKS
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NanoVNA and TinySA for Radio Design

- 1. NanoVNA and TinySA for Radio Design (18:57)
- 2. NanoVNA - Overview and antenna measurements with S11 (9:48)
- 3. NanoVNA Demonstrations - Coax line reflections and Smith charts (20:26)
- 4. NanoVNA - Measuring RLC Components (30:16)
- 5. NanoVNA - Measuring Impedances (14:03)
- 6. NanoVNA as a synthesized CW signal generator (16:15)
- 7. NanoVNA Calibration - When, Why, and How to cal a VNA (6:20)
- 8. NanoVNA - Antennas and Tuners (29:59)
- 9. NanoVNA - Measuring S21 and S11 of a small-signal amplifier (5:35)

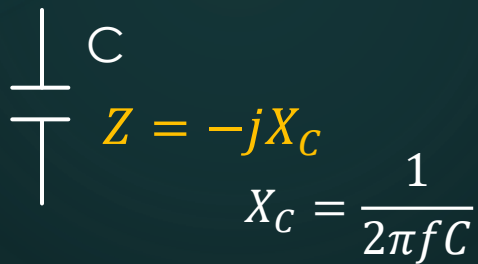
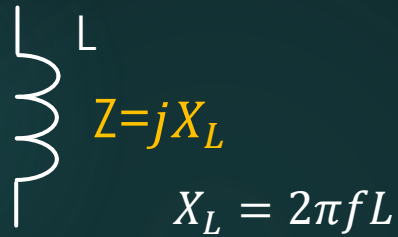
ECE Topics
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Topics in ECE #1
Voltage, Current, Resistance, and Power - ECE Topics #1

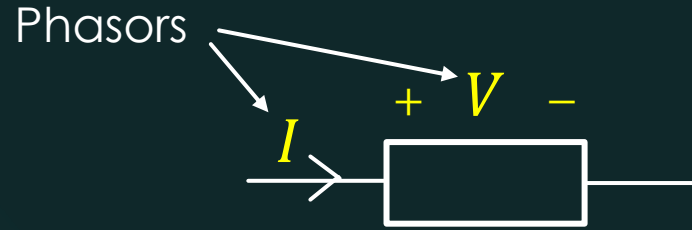
- 1. Voltage, Current, Resistance, and Power - ECE Topics #1 (27:13)
- 2. Power, Energy, and Electric Vehicles - ECE Topics #2 (30:34)
- 3. Circuit Analysis - From Theory to Applications - ECE Topics #3 (35:33)
- 4. Understanding Circuits - ECE topics #4 (part 1) (26:05)
- 5. Understanding Circuits - Part 2 - ECE Topics #4A (41:19)
- 6. Oscilloscopes, Microphones, and Noise - ECE Topics #5 (14:05)
- 7. Complex Numbers, Phasors, Impedances, and Frequency Response (48:28)
- 8. Frequencies, Amplitudes, Log Scales, dB and dBm - ECE Topics #7 (36:15)
- 9. Fourier and FFT Concepts in Circuit Design - Part 1 (ECE Topics #8) (23:38)

Associated website: <https://ecefiles.org/>

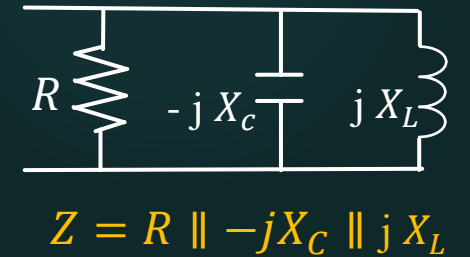
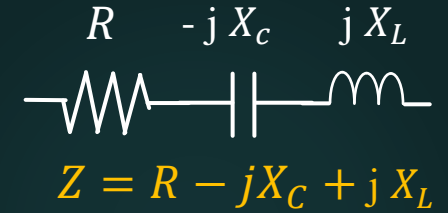
R L C Impedances



$$Z = R + jX$$



$$V = I Z \quad \text{or} \quad I = \frac{V}{Z}$$



$$f_o = \frac{1}{2\pi\sqrt{LC}} \quad BW = \frac{f_o}{Q}$$

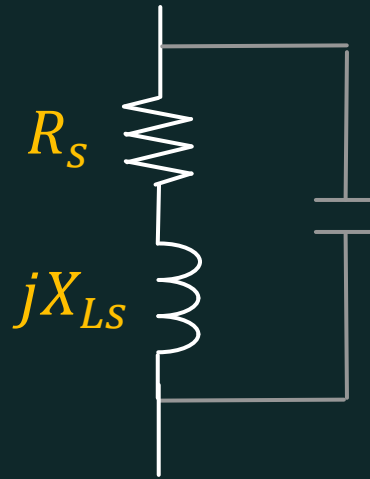
$$Q_s = \frac{X_s}{R_s} \quad Q_p = \frac{R_p}{X_p}$$

Real-World Components at RF

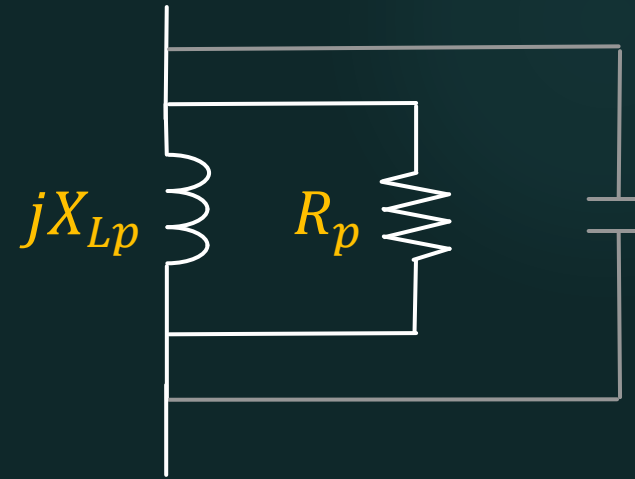
Ideal



Real-World (Simplified Models)



Series
Loss Resistance

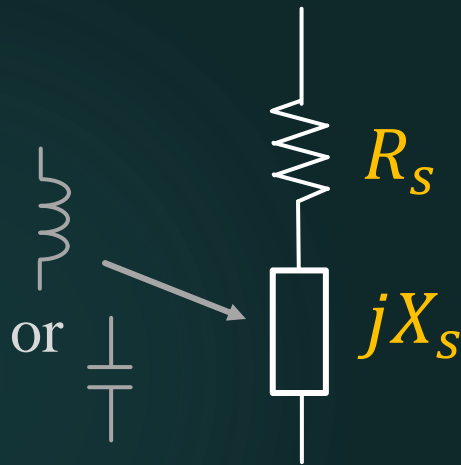


Parallel
Loss Resistance

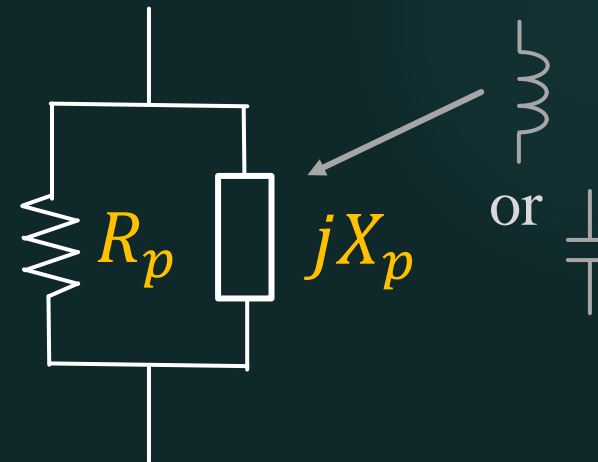
Theorem

At a specific frequency, an impedance $Z = R + jX$ can be realized in a series form or a parallel form:

Series form:



Parallel form:



NOTE: At the specific frequency (say f_o), these two realizations are *equivalent* – they will yield the same circuit behavior and can be swapped to make circuit analysis/design/production easier/cheaper ...

(Constructive) Proof

Series form is trivial: $R + jX = R_s + jX_s$

Parallel form is messy: $R + jX = R_p \parallel jX_p$

$$R_p \parallel jX_p = \frac{(R_p)(jX_p)}{(R_p + jX_p)} = \frac{(R_p)(jX_p)(R_p - jX_p)}{(R_p + jX_p)(R_p - jX_p)} = \frac{R_p^2(jX_p) + R_p(X_p^2)}{R_p^2 + X_p^2} = R_p \frac{X_p^2}{R_p^2 + X_p^2} + jX_p \frac{R_p^2}{R_p^2 + X_p^2}$$

So, $R + jX = R_s + jX_s = R_p \frac{X_p^2}{R_p^2 + X_p^2} + jX_p \frac{R_p^2}{R_p^2 + X_p^2}$

And, hence...

$$R_s = R_p \frac{X_p^2}{R_p^2 + X_p^2} \quad X_s = X_p \frac{R_p^2}{R_p^2 + X_p^2}$$

Cleaning up (and solving for R_p, X_p)

Define a quality factor $q = R_p / X_p$

Then

$$R_s = R_p \frac{X_p^2}{R_p^2 + X_p^2} \left(\frac{1/X_p^2}{1/X_p^2} \right) = R_p \frac{1}{q^2 + 1} \quad \text{And} \quad X_s = X_p \frac{R_p^2}{R_p^2 + X_p^2} \left(\frac{1/X_p^2}{1/X_p^2} \right) = X_p \frac{q^2}{q^2 + 1}$$

Re-arranging yields ...

$$R_p = (1 + q^2) R_s$$

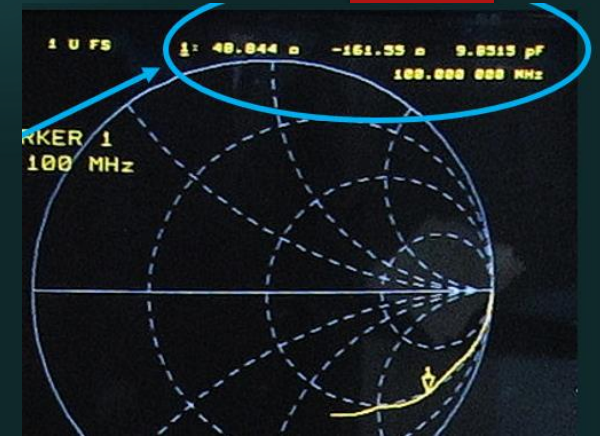
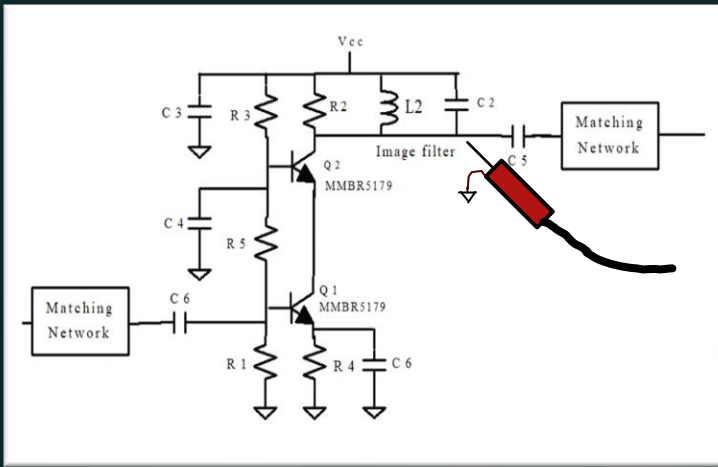
$$X_p = \frac{q^2 + 1}{q^2} X_s = \left(1 + \frac{1}{q^2} \right) X_s$$

NOTE: It can also be shown that

$$q = X_s / R_s$$

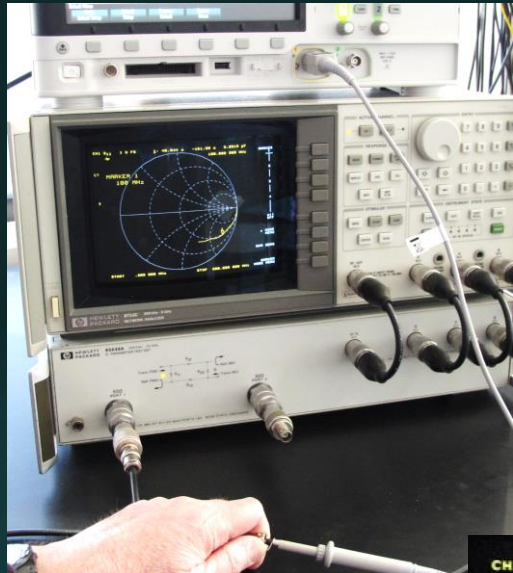
The proof is left as an exercise for the reader ;-)

Applications



1. Convert series-form simulation or measurement results to parallel form, to better understand effects on circuit
2. Derive / understand impedance matching networks
3. Quantify real-world inductor and capacitor parasitics
4. Design RF amplifiers !

Convert Series Z to Parallel Form



Measured loading of probe at 100 MHz in series form is:

$$Z_{meas} = 48 - j 161$$

$$q = \frac{161}{48} = 3.35$$

$$R_p = (1 + 3.35^2)48 = 588! \quad (\text{not } 10 M)$$

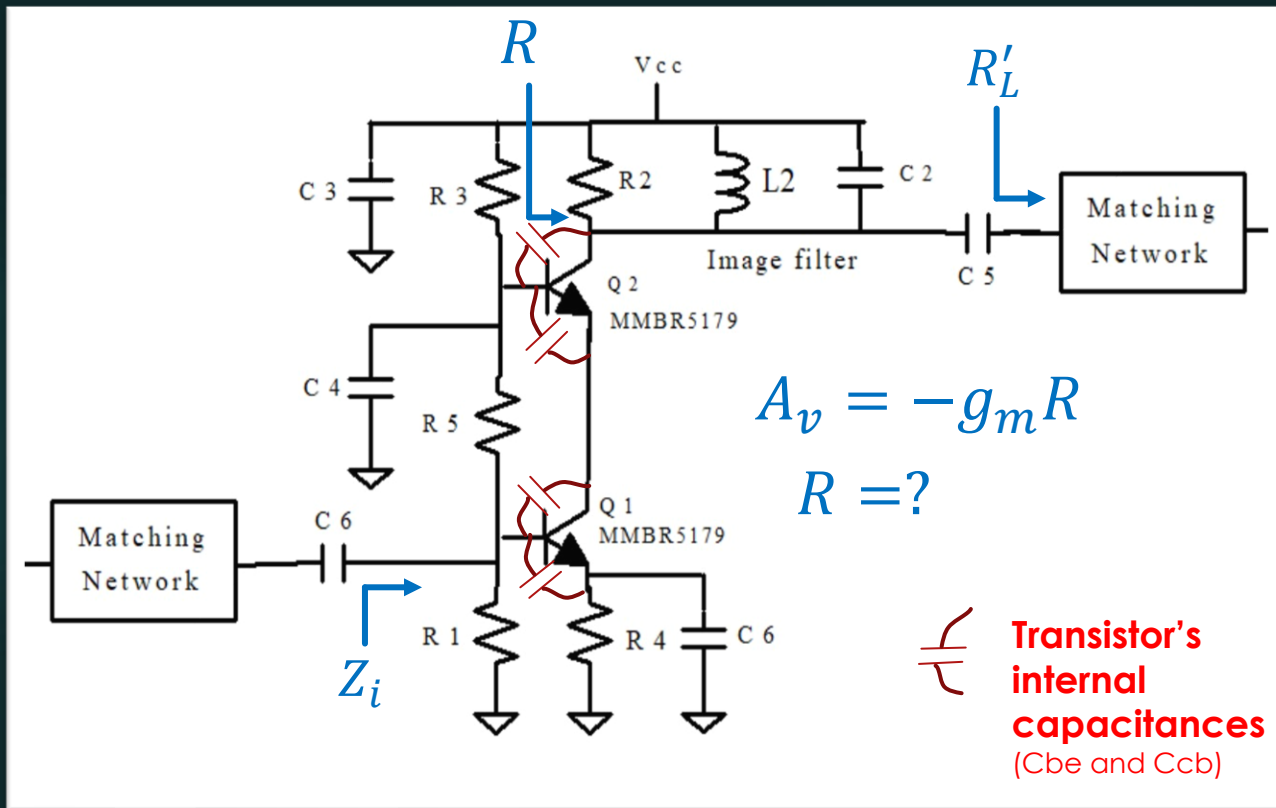
$$X_p = \left(1 + \frac{1}{3.35^2}\right) 161 = 175$$

$$C_p = \frac{1}{2\pi f X_p} = 9.1 pF \quad (\text{a bit less than } 15p)$$

What is
 $48 - j 161$
in parallel form ?



Real-World Inductors



Assume $L_2, C_2 + C_{cb}$ resonate at 100 MHz,

Suppose $R_2 = 3K$ and $g_m = 20mA/V$

Then, for matched load ($R'_L = 3K$), we should have $R = 1.5K$ and $A_v = -30 V/V$ (29.5 dB)

if L_2 is ideal

Practical inductors have Q from 10 to 100, depending on size, construction.

Suppose L_2 is 100 nH ($X_L = 63 \text{ Ohms @ } 100 \text{ MHz}$)

If Q of L_2 is 10, then

$$R_s = 6.3, \text{ and}$$

$$R_p = (1 + 10^2)6.3 = 636 \text{ Ohms}$$

and R falls to 447 and A_v falls to 8.9 V/V (19 dB) !

L Matching Network Design

3

Summary

$$\frac{R_s}{R_p} \frac{jX_s}{jX_p} = \frac{R_s}{R_p} \frac{jX_s}{jX_p}$$

Alternative Formulation

Define $g = \frac{X_s}{R_s}$

Then $R_p = (1+g^2) R_s$ $X_p = \frac{1+g^2}{g} X_s$

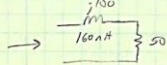
Can also show: $\frac{R_p}{X_p} = g$

"Compare of previous formulae for $Q_p, Q_s!$ "

Example:

Consider $Z = 50 + j100 \Omega$ at 100 MHz

Series Resistor

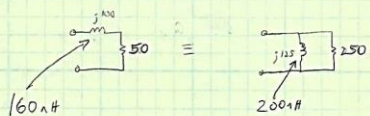


Parallel equivalent (at 100 MHz)

$$g = \frac{X_s}{R_s} = \frac{100}{50} = 2$$

$$R_p = (1+g^2) R_s = (5)(50) = 250 \Omega$$

$$X_p = \frac{1+g^2}{g} X_s = \frac{5}{2} 100 = 250 \Omega$$



4

HANDOUT

Z matching with L networks

Example

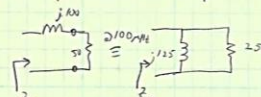
Convert 50Ω to 250Ω at 100 MHz

Step 1
Add series inductor



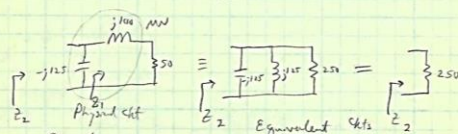
Step 2

Views on parallel form (see prev. example)



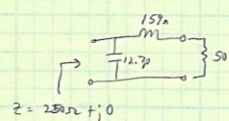
Step 3

Resultant parallel equivalent reactance



Step 4

Convert reactances to L, C values



$$Z = 250 \Omega + j0$$

@ f_0 only

$$X_L = 2\pi f L \Rightarrow L = \frac{X_L}{2\pi f} = \frac{100}{2\pi(10^8)} = 159 \text{ nH}$$

$$X_C = \frac{1}{2\pi f C} \Rightarrow C = \frac{1}{2\pi f X_C} = \frac{1}{2\pi(10^8)(12.5)} = 12.7 \text{ pF}$$

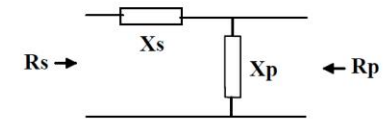
Step 5 check for practicality (PCB load)

$$C \geq 1 \text{ pF} \quad \checkmark$$

$$L \geq 10 \text{ nH} \quad \checkmark$$

L Matching Network Design

Basic L network:



Design Procedure:

- Given a desired transformation $R1 \leftrightarrow R2$, equate the lowest resistance (call it $R1$) with the "series side" of the matching network. I.e. $R1 = R_s$ and $R2 = R_p$.

- Compute the required q value from:

$$q = \sqrt{\frac{R_p}{R_s} - 1}$$

- Compute the series reactance from:*

$$X_s = \pm q R_s$$

- Compute the parallel reactance from:*

$$X_p = \mp \frac{1+q^2}{q^2} X_s = \mp \frac{R_p}{q}$$

- Convert to capacitance and inductance values using

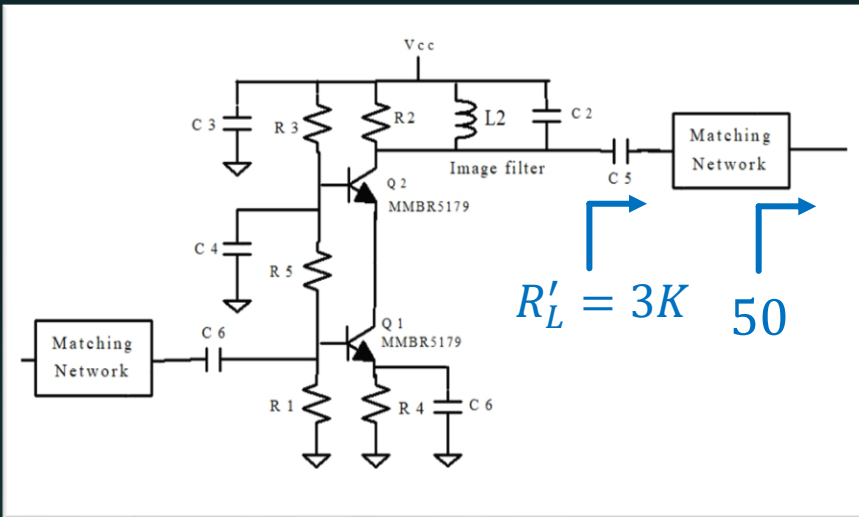
$$L = \frac{X_L}{2\pi f} \quad \text{and} \quad C = \frac{1}{2\pi f X_C}$$

* If one component is selected to be capacitive, then the other must be inductive. The choice of which type to use for a particular side of the network depends on other circuit design considerations. For example, the choice:

- will determine whether a "lowpass" or "highpass" response is present outside the primary matching frequency,
- may affect the bias circuit design, or
- may make it possible to save on the number of circuit components by careful combinations of the functions of components such as the matching capacitor and a required AC output coupling capacitor.

Matching Network Example

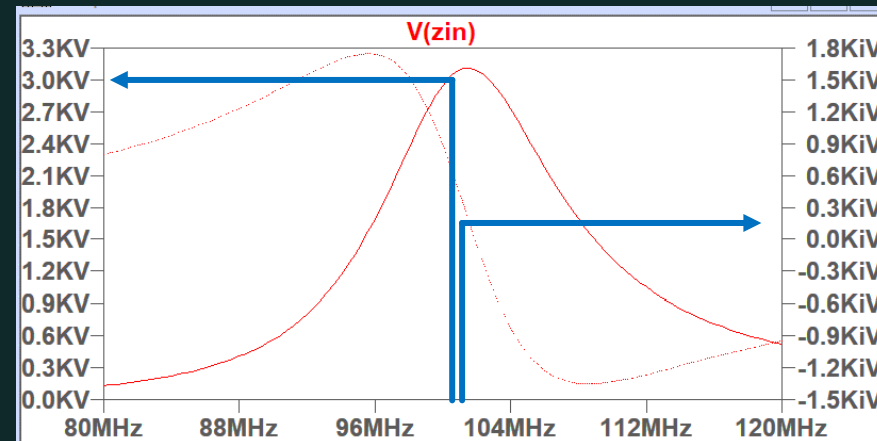
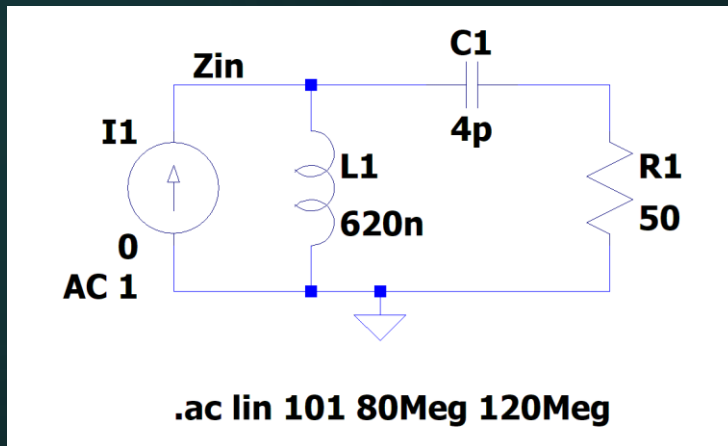
Design/verify output MN for converting 50 Ohm load to $R'_L = 3K$ at 100 MHz



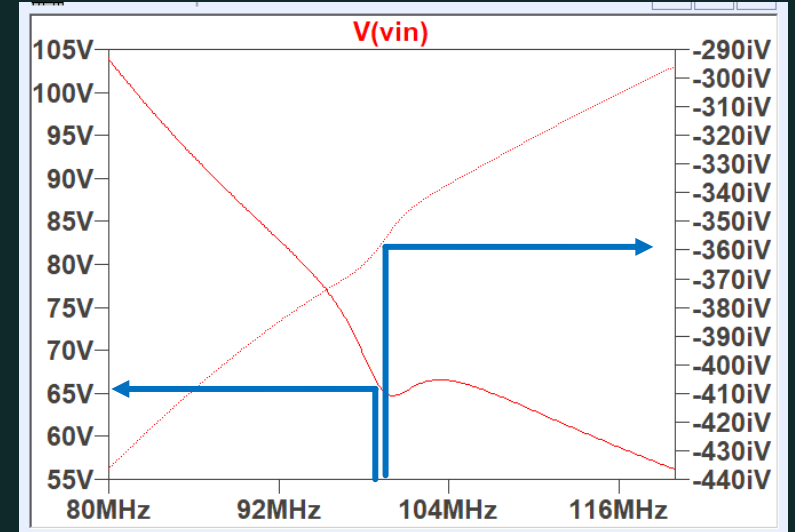
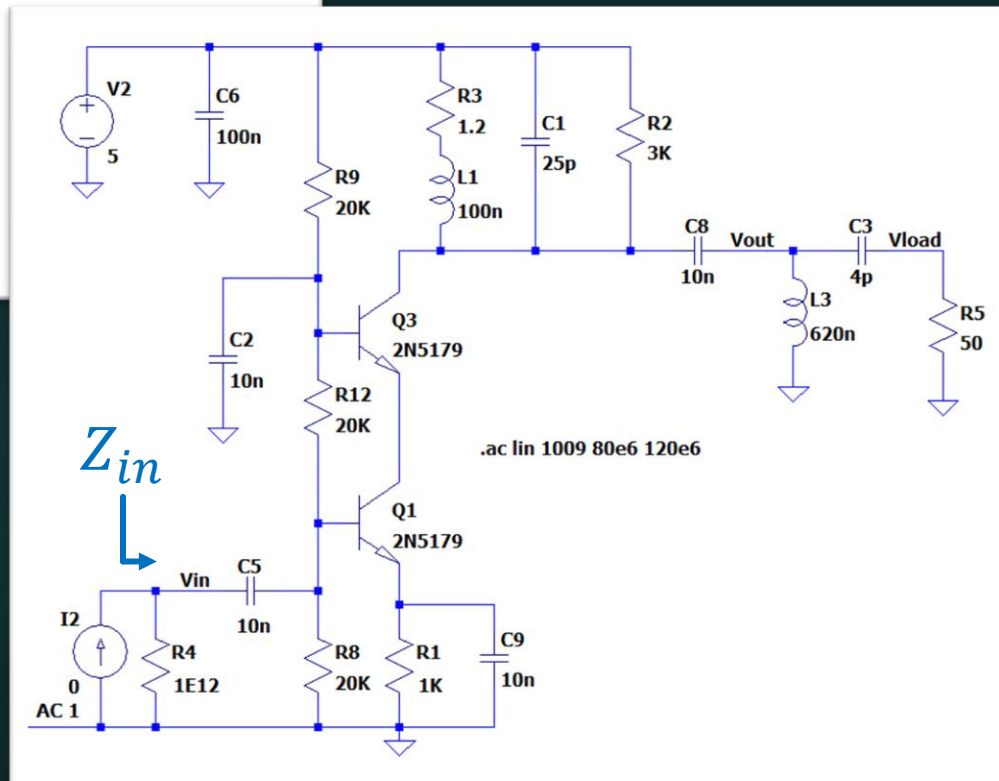
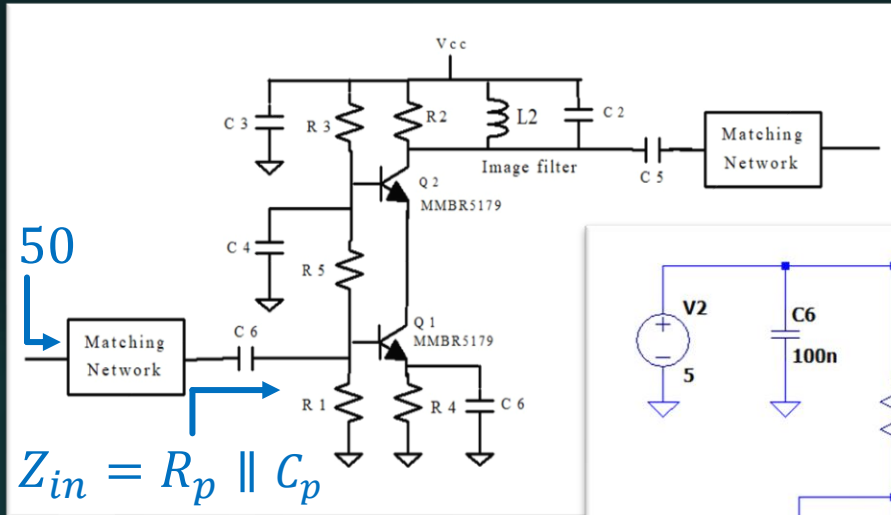
$$q = \sqrt{\frac{R_p}{R_s} - 1} = \sqrt{\frac{3K}{50} - 1} = 7.68$$

$$X_s = q(50) = 384 \Rightarrow C_s = \frac{1}{2\pi(100E6)(384)} = 4.15pF$$

$$X_p = (3K)/q = 391 \Rightarrow L_p = \frac{391}{2\pi(100E6)} = 622nH$$



Matching to Complex Z_{in}



At 100 MHz,

$$Z_{in} = R_s + j X_s = 65 - j 360$$

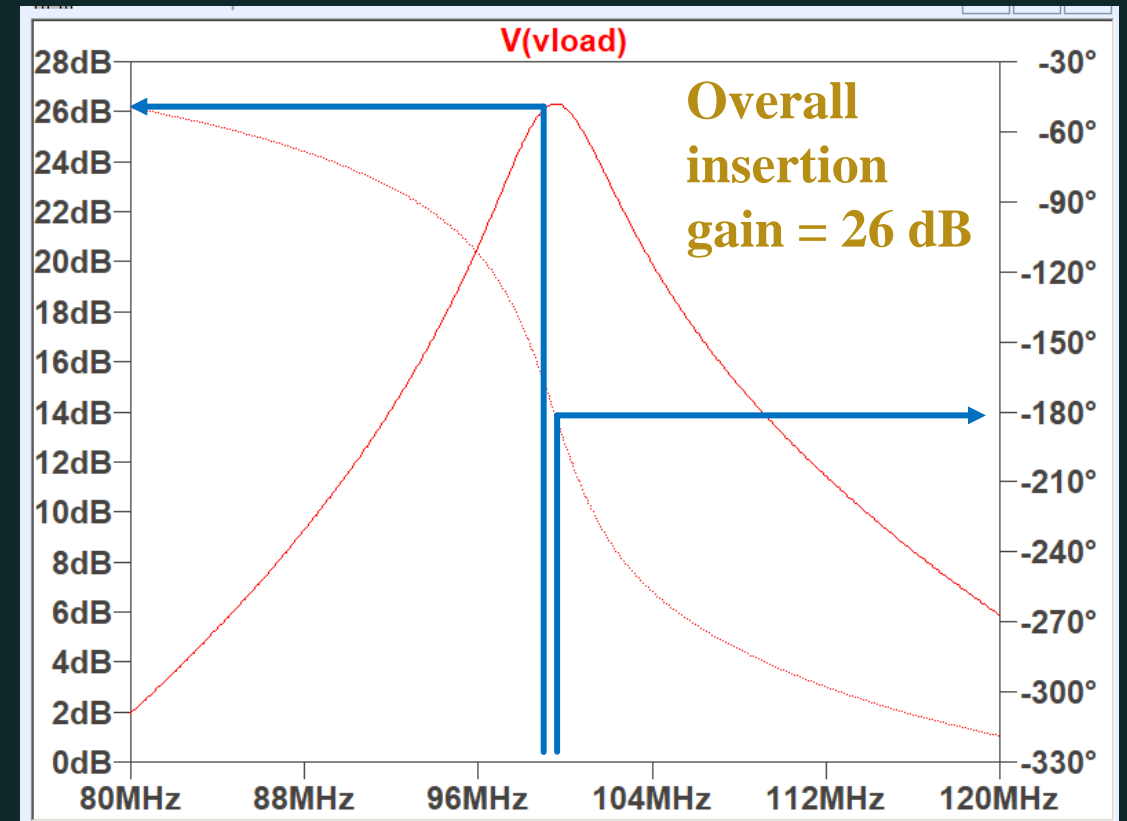
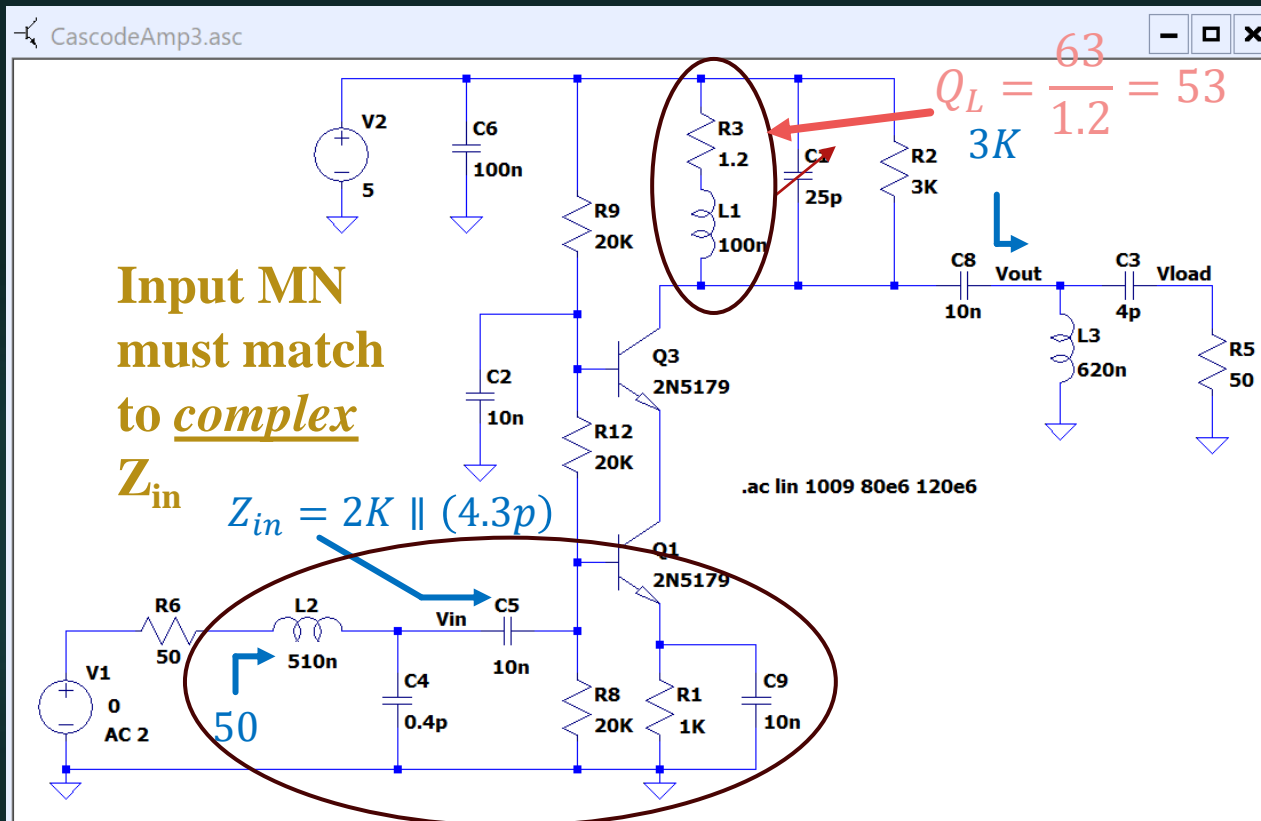
$$q = \frac{360}{65} = 5.54$$

$$R_p = (1 + q^2) R_s = 2.06K$$

$$X_p = \frac{(1+q^2)}{q^2} X_s = 372$$

$$C_p = \frac{1}{2\pi(100E6)372} = 4.28 \text{ pF}$$

Full RF Amp Simulation 😊



NOTE: We should probably delete R2 here. Why ?

For More Information

Radio Design 101
MegawattKS
13 videos • 46,321 views • Last updated on Nov 22, 2023

Play all Shuffle

A collection of videos abstracted from a university course on radio / RF circuit design. The goal is a working FM broadcast receiver, but the material is applicable to all wireless hardware from Amateur Radio to commercial RF integrated circuits. While the focus is on circuit design, episodes in this series touch on a large set of topics ranging from basic circuit and system architectures to radio performance measurement and optimization. More information on the videos and the university course from which it came is available at: <https://ecefiles.org/>

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MegawattKS • 12K views • 2 years ago
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NanoVNA and Radio Frequency / Microwave Tech
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Educational videos pertaining to the amazing NanoVNA and TinySA products, and to the radio frequency (RF) and microwave technology underlying these low-cost, high performance instruments.

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This evolving series on Electrical and Computer Engineering (ECE) Topics is based on and designed to support university-level introductory courses as well as independent study and review.

Prerequisites are generally limited to algebra and geometry, although we introduce some more advanced concepts like complex numbers when needed. But the series concentrates on concepts rather than math(ε).

Throughout the series we use real-world products as examples to relate theory to practice and hopefully motivate further study. We hope you find these videos helpful.

For those interested in radio frequency / wireless electronics, please see our other playlists including

"Radio Design 101", "Antennas and Propagation", and "NanoVNA and Radio Frequency/Microwave"

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